

IMPROVING THE EFFICIENCY OF MULTI-COLOR LEDS DISPLAY SYSTEM USING THE IMPROVED PWM TECHNIQUE ON FPGA.

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ABSTRACT

Light Emitting diode (LED) video displays are future of large-area display market. Their scale, their durability and their ability to produce an attractive video in ambient light conditions, make them the favorable screens in all display systems. To increase the light quality of LED displays, it is necessary to improve the techniques that control the light variations while keeping stable energy consumption. This technique has become a major objective in the design of such displays systems. This paper presents a comparative study between two techniques to control brightness levels of LED displays, the first technique controls by varying direct current amplitude (DC), the second by varying the duty cycle of the current square wave with fixed amplitude (PWM). The paper also proposes an improved PWM solution, which is based on the approximation of the PWM curve to the DC curve in order to have the luminous efficiency as the last one. The demonstration is presented as a serial of mathematical calculations to choose the intermediate point and build low-cost hardware algorithm for this new method. The performance of this method is studied experimentally using the FPGA design and is compared with the PWM method.

Keywords: LED, Display systems, Display driver, PWM, FPGA, Grayscale, multiplexing LEDs,

1. INTRODUCTION

Developments, industrial and commercial, have increased exponentially in recent years [1]. With this development, the technology has been able to facilitate multiple services by exploiting the different types of energy sources, but on the other hand this development increased the energy consumption in the world. Based on data from the US Department of Energy, approximately 30% of the electric power has been consumed by the commercial and industrial lighting [2]. Reducing the power dissipated by the lighting is then a major factor in saving energy. Therefore, studies on improving these systems and the lighting energy savings deserve more attention from academic researchers.

LED screens have large big of the market for large area display. Their scale, their durability, their ability to produce an attractive video and their low power consumption, make this technology favorable than other display systems. They are using to the outdoor advertising across a

wide range of applications including stadiums, arenas, a complex of entertainment, etc [3].

Using PWM technique to control the luminous of RGB LED and reduce energy consumed using DC technique (converter digital to analog) in these large-area displays, caused an error gap between color levels produced by these two techniques. The main objective in the design of display systems is to improve luminous efficiency by reducing this error gap, maintaining stable the cost and the energy consumption of these systems. For this, the methods of controlling the brightness of each LED in the display matrix still needs improvement in order to increase the light quality without instability in energy consumption in the entire system.

Several studies have been done on these methods; in general, they can be divided into two main categories [4]: (1) Analog or direct current (DC) using the D/A converters and (2) Pulse-width modulation (PWM). There are many studies have reported the influence of the variation of direct current on the luminous efficiency RGB LEDs.

On other hand, the optical characteristics of RGB LEDs on this display system remain unknown.

The practical uses luminous control techniques for each component R, G or B of RGB LEDs still need improvement.

This paper presents an efficient design of an RGB LEDs display system controller by improving the performance of the luminous dimming method of each RGB color component. This design replaces variation of direct current amplitude DC by variation PWM while changing its characteristic (current-luminous) by increasing its variation range.

This paper is organized as follows; the first section introduces the LED display structure and luminous control methods of each LED. The second section describes the approach of the improved PWM method and the performance compared to DC and simple PWM methods. Simulation and experimental results are presented and discussed in section 3. Finally, the paper ends with a conclusion.

2. CONTROL METHODS OF COLOR LUMINOUS.

LED multicolor large-area display systems require a complex treatment that requires high performance conditions. Using the multiplexing technique (Fig. 1), the display driver can manage a large number of tri-color LEDs RGB organized in the form of matrices.

This method allows to scan all the LEDs and assigning to each color component (r, g and b) the level of corresponding brightness. This operation requires a very high speed in the order of KHz so that the human eye cannot distinguish the fluctuations of the image [5]. Therefore, it is very important to design an efficient display driver. In particular, the efficient method of assigning information of each pixel to the corresponding LED (part bordered by the green color in Fig. 1). This operation affects directly the energy efficiency of the display as well as the quality of the displayed image. In this sense, several studies have been done on these methods [6] [7] [8]. In general, they can be divided into two main categories:

- Analog or direct current (DC).
- Pulse width modulation (PWM).
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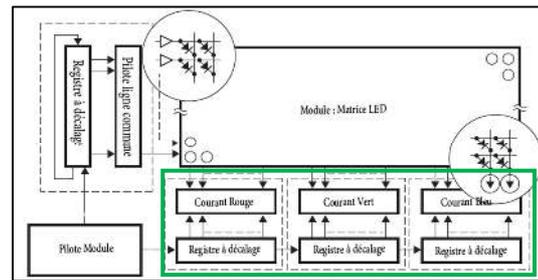


Fig. 1 : The multiplexing method of an LED matrix RGB.

The first method controls luminous by varying the amplitude of a DC current (AM variation technique). It can be obtained based on the principle of the digital to analog converters. This principle is done by varying a variable resistor in order to affect a current level corresponds to the desired illumination of the LED [9]. The analog control provides higher luminous efficiency and stability of the color variation [8].

However, this method has two weaknesses first one energy consumed by these converters cause the interference electromagnetic, secondly the intrinsic optical properties of the LEDs which the output varies non-linearly with the DC current variation, this method is suitable for lighting applications that do not require linear variation [10].

a. Pulse Width Modulation Algorithm

The problem associated with the DC method is satisfied when the PWM method is used. By this approach, the energy consumed by the analog digital converters is preserved by replacing these DACs by a constant direct current in the form of a square signal of period T and duty cycle t/T . By exploiting the very fast response time of the LEDs which can be controlled by using succession of current pulses, This current flows through the LEDs for a duration of the pulse '1' (tFig. 2), and completely extinguishes them for the duration ($T-t$ in Fig. 2). For example, the current pulse with a period T of 1s (frequency 1Hz) with a duty cycle of 50% and current with amplitude of 2A has the same effect as a driving current of 1A in terms of brightness [9] (Fig. 2).

By this method, a chopped direct current is produced. The LEDs can be attenuated by varying the duration t of the pulse '1', which generates an output with different levels of luminance proportional to the PWM current I , as illustrated Fig. 3.

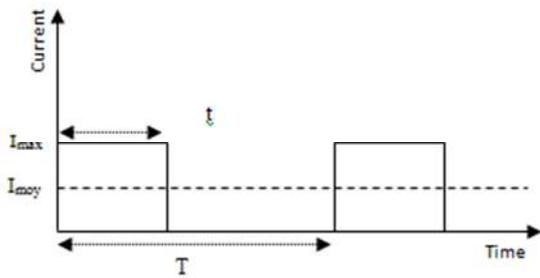


Fig. 2 : Current Signal Formats for PWM Technique [10].

In addition to the linear variation, the PWM method is also intrinsically compatible with the operation of the display systems. The LED Luminous is controlled directly from the brightness information embedded in the digital input signal. This information can be corresponding to the time t of the Pulse '1' of the PWM signal as showing in Fig. 2.

The disadvantage of the PWM method is that it produces a reduced luminous of the LEDs than DC method, for example the luminance PWM for a current value I_1 is less than that of DC (Fig. 3).

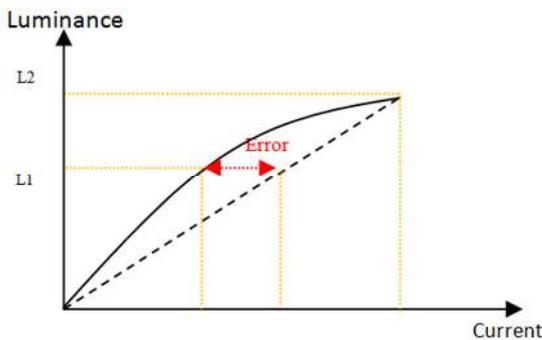


Fig. 3 : The curves associated with two methods of variation of the luminous intensity of the LEDs.

b. Hardware Description of PWM

In the driver design process, there are two methods for performing the PWM technique: the counter comparison method [9] and the method of separating bit planes according to weight. The first method generates an initial value of 0 and its maximum value is 255, the counter continues counting according to the clock cycles. The initial state of the LED is set to '1'. During the next clock cycle, the data (0-255) read is compared with the counter data, if the read data is greater than the value of the counter, the conduction time of the pixels is not sufficient.

Therefore the LED must remain in state '1', if the data is less than the value of the counter, the

conduction time of pixels complies with the requirements. Therefore the LED goes from state '1' State '0'. In practice, for this approach to read the data and compare it 256 times for each frame and if the screen is slightly larger, it is necessary to have a very fast speed to have a flicker-free display [11][12], which is a disadvantage of this method.

For the second method, only one bit of the digital signal is displayed among the eight data bits (D7 - D0), and the signal can be displayed 8 times to present all eight bits of data except that the bits must take the different durations according to the weight of each of them as illustrated in Table 1.

The corresponding light intensity is the sum of the bits which have the state 1.

Table 1 : The current intensity variation by the PWM method.

Bit of data	Duty cycle (t/T)
8 th Bit	128/255
7 th Bit	64/255
6 th Bit	32/255
5 th Bit	16/255
4 th Bit	8/255
3 th Bit	4/255
2 nd Bit	2/255
1 st Bit	1/255

At RGB component data of each pixel, for each bit '1' corresponds a cyclic ratio t/T according to their weights ($D_7, D_6, D_5, D_4, D_3, D_2, D_1, \text{ and } D_0$). The conduction time given to the low luminosity is short, while the conduction time given to the high luminosity is long. Thus, the full cyclic ratio of the 8-bit data can be calculated by the sum of the cyclic ratios of the values of '1'.

c. Display control adapted to the PWM method

A typical LED driving module, as shown in Fig. 4, contains a 16-bit shift register which receives data transmitted serially by the FPGA controller. The FPGA circuit is programmed to retrieve 16 data bits and transmit them to the 16-clock LED driver module. LED display modules that contain more than 16 pixels per line require multiple registers to be cascaded into the driver. LED display blocks that contain more modules (also more pixels per line) require the cascading of multiple module drivers.

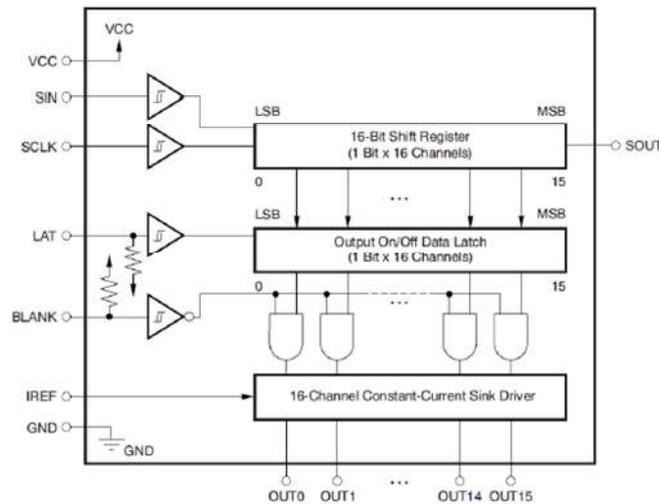


Fig. 4 : Typical LED driver module [13].

The FPGA circuit continues to transmit serial data until all shift registers are populated with data bits (D_i). When the transmission of all the data bits corresponding to D_i is completed, the signal LAT will be activated to lock the transmission of the data bits and to pass to the next data bits D_{i+1} . The current-constant LED drivers are then activated by lowering the BLANK signal for a duration corresponding to the weight of the stored data bits. The same process is repeated until all eight groups of data bits (representing the eight different weights) are transmitted to the LED driver modules. After this value input to the row decoder, the counter corresponding to the ABCD inputs is incremented by 1 to select the next row, transmission of the data bits for that row will begin and the LEDs of the pixels will be activated in a manner similar to the first one.

A typical example of a timing diagram corresponding to this operation is shown in Fig. 5.

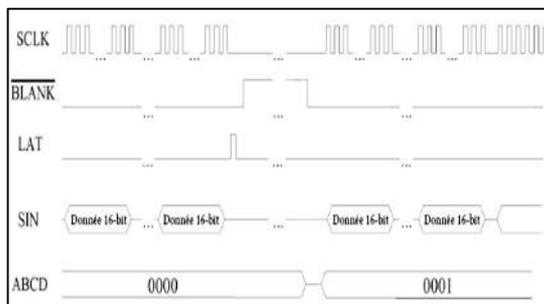


Fig. 5 : Timing diagram showing the various control signals applied to the LED control module.

3. PROPOSING METHOD TO IMPROVE PWM

To take account of the non-linearity introduced by the DC method and to ensure that the color quality is preserved, additional calculations are needed to estimate the mean PWM current required to

produce the same luminance as the originally DC current. These simple calculations are based on the slopes m_1 and m_2 of the two linear portions of each section (ON) and (NM) as shown in figure Fig. 6. Consider the continuous variation curves of an LED, which can generally be approximated by equation (1) to a quadratic form.

$$(1) \quad y = -a * x^2 + b * x$$

Where 'x' and 'y' represent respectively duty cycle (coded on 8 bits) and standardized luminosity ($y=L/L_{max}$) equivalent to the applied technique. The coefficients 'a' and 'b' can be obtained by performing a curve fitting to the DC gradation obtained from an experiment (example next section).

Assuming that the original signal has a PWM duty cycle of 'x', the normalized luminance 'y' corresponding to the PWM technique is expressed by equation (2):

$$(2) \quad y = x / 255$$

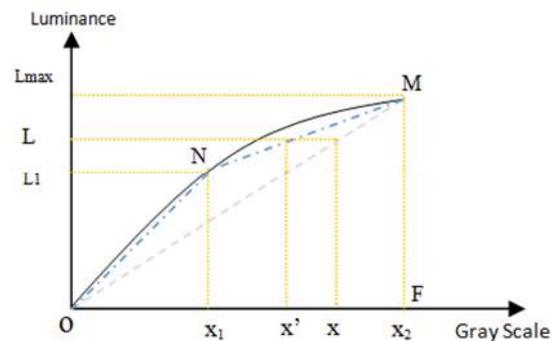


Fig. 6 : Amélioration de la méthode PWM.

Referring to figure Fig. 6, and if we consider x_2 corresponds to the large value of x ($x_2=255$). The duty cycle PWM equivalent to the improved PWM current is denoted by x' . It can also be seen in this figure that if $L < L_1$ the required improved

PWM duty cycle value is in the lower part of the linear section (ON) with a slope m_1 . Then the current PWM will have a duty cycle PWM equivalent to x' calculated by the equation (3).

$$(3) \quad x' = \frac{x}{255 * m_1}$$

If $L \geq L_1$, the linear section is in the upper part (line (NM) with slope m_2). The PWM duty cycle equivalent to their current can be calculated from equation (4).

$$(4) \quad x' = \frac{1}{m_2} * \left(\frac{x}{255} - y_1 \right) + x_1$$

Using the DC curve approach, conversion can be done more easily using equation (5), assuming that the DC variation curve has been approximated by the two combined linear sections.

$$(5) \quad x' = \frac{b}{2*a} - \frac{1}{2*a} * \sqrt{b^2 - \left(\frac{2*a}{255} \right) * x}$$

a. The intermediate point of luminous efficiency $I_1(x_1, y_1)$.

The concept of "luminous efficiency" is used here to reflect the fact that the light emitted by LED continuously changes during the display of the images and it is assumed that the LEDs have the same probability of operating at any point on the variation curve.

As illustrated in la Fig. 6, the task is equivalent to maximizing the area between points O, N, M and F (ONMF). This surface is calculated by equation (6) if we take y_2 is maximum luminance which corresponds to x_2 in point M:

$$(6) \quad \text{ONMF} = \frac{1}{2} * y_1 * x_2 + \frac{1}{2} * y_2 * (x_2 - x_1)$$

The values y_1 and y_2 are replaced by their corresponding values according to equation (1):

$$(7) \quad \text{ONMF} = \frac{1}{2} * a * (x_2^2 * x_1 - x_2^3 - x_1^2 * I_2) + \frac{1}{2} * b * x_2^2$$

To determine the x_1 which maximizes the ONMF area, the derivative of ONMF with respect to x_1 is set to 0, which gives equation (8):

$$(8) \quad x_1 = + \frac{1}{2} * x_2$$

The result is independent of the coefficients of the quadratic function; it is also applicable to all types of LEDs which have a luminous output characteristic.

In fact, the result remains valid even for the case of $a=0$, rotating the quadratic function in a linear

function $y = b * x$, in which case the improved PWM method will have the same performance as the PWM method, There will be no gain in luminous efficiency.

b. The hardware description of improved MPW algorithm.

To achieve the improved PWM method on LED display systems based on the module shown in Fig. 4, the reference current I_{Ref} can be divided to two same intensity current I_{NRef} (with $I_{NRef} = I_{Ref}/2$). The I_{NRef} can be used as an intermediate point between 0 and $I_{Ref} = 2 * I_{NRef}$ as shown in Fig. 7. Affection of both of these currents will be controlled directly by the FPGA driver. Indeed, the first driver is configured to deliver a maximum current of I_{NRef} to the LED pixels. Therefore, the use of second parallel current provides a maximum current of $I_2 = 2 * I_{Ref}$ to be delivered to the LED pixels.

To control with this method, the FPGA driver separate the originally data (8-bits of pixel) to two buses for controlling the value of each I_{NRef} . For example, considering that the required intensity level of a given pixel component is represented by the 8-bit signal: [10100000]. If a single LED control module with I_{Ref} is used, the average current I_p delivered to the LED pixel will be calculated according to the following equations:

$$(9) \quad I_p = \left(\frac{255}{255} + \frac{65}{255} \right) * I_{NRef}$$

$$(10) \quad I_p = I_{NRef} + \frac{65}{255} * I_{NRef}$$

Equation (10) means that the same average current the 8-bit signal: [1010 0000] can be obtained using direct constant current I_{NRef} and LED driver module (with I_{NRef}). The first equivalent to $255/255 = [11111111]$, and the other with the binary equivalent of $65/255 = [01000001]$.

Based on the above example, if 'x' is equal to the decimal equivalent of the 8-bit signal for a pixel component, the equivalent signal can be converted into two sub-signals 8-bit consisting of the following:

If $x \geq 128$: the I_p equal to I_{NRef} plus [binary equivalent of $2*x-255$].

If $x < 128$: the I_p equal to [binary equivalent of $2*x$].

$x = 128$ corresponds to the original 8-bit signal having the MSB=1, the MSB value can be used to distinguish between two possible cases set forth above. So this conversion can be performed very efficiently without repeating the calculation steps given in equation (9).

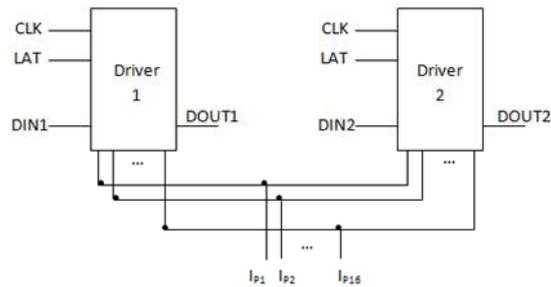


Fig. 7 : Using the paralleling of two drivers to have three levels of currents.

The proposed structure can reach more than two current levels, if this method finds recognition and adoption on the part of the LED display industry in the future. Integrated circuits containing several control units can be manufactured to achieve better brightness with only a small increase in the cost of the equipment.

4. EXPERIENCE, RESULTS AND DISCUSSION

In order to demonstrate the feasibility of the algorithm proposed, an experiment was carried out as a hardware architecture using Xilinx’s ISE (Integrated Software Environment), synthesized on a single Xilinx’s and implemented in Nexys2 Digilent card based on the spartan3e XC3S500E

FG320 FPGA circuit which contains about 500k logic gates and which can be manipulated with a processing speed of 50 MHz adapted to the design of LED display system [14]. As shown in Fig. 8, this card contains several communication devices (PS / 32, Hi-speed, Pmod, VGA port, RS232, FX2_IO...) that surround the FPGA circuit.

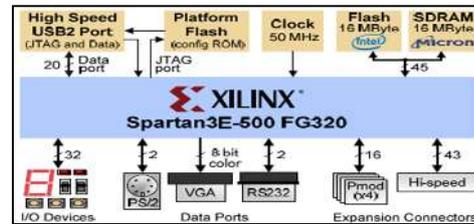


Fig. 8 : Constitutions of the Digilent Nexys2 Card

The Fig. 9 shows the data processing and correction block, this block contains two sub-blocks; the first block treats input data using the algorithm explained above then it is separated to two sub-signals *pixel_pwm1* and *pixel_pwm2*. These signals are sent separately to two coders blocks *Codeur_PWM*. They take the decision and assign different levels to the two outputs signals (r_1, g_1, b_1) and (r_2, g_2, b_2).

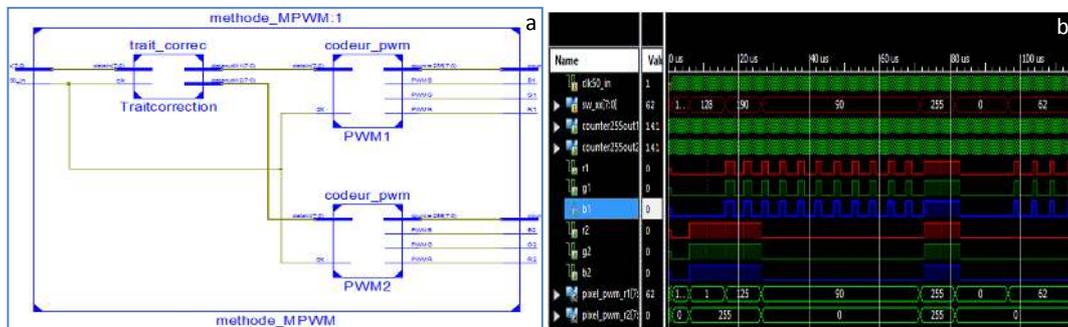


Fig. 9 : (a) Block test data processing and correction. (b) Simulation of the block processing test and data correction

Firstly, the block test of improved PWM method was simulated to check behavioral operating. The

Fig. 11 shows behavioral simulation results in the form of timing diagram obtained by the hardware simulation using ISim-HDL of the above processing block, it shows the timing of each main signals. The pixel data signal was presented by input *sw_xx* (0, 90, 190 and 255), on the

Fig. 11 when *sw_xx* equal to 90 signal *pixel_pwm1* has same value and signal *pixel_pwm2* turn to 0, however for *sw-xx* equal to 190 signal *pixel_pwm1* equal to (2*190-255=125) and *pixel_pwm2* equal to 255... ect. Next, we evaluated the performance of improved PWM implemented on FPGA using the eight

switches to input pixel data (0 to 255) and the six fast FX2_IO to generate r_1, g_1, v_1 and r_2, g_2, v_2 outputs to LED. as shown in Fig. 10.

In order to evaluate the output results, we are using oscilloscope to visualize the variation of the output signal according to the value entered by the switches. .

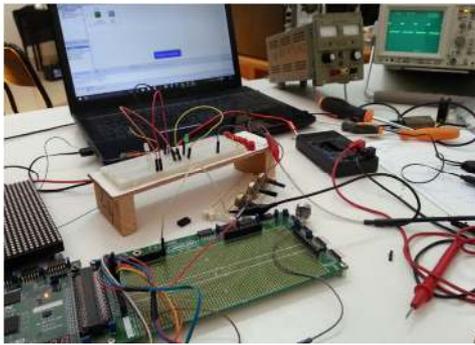


Fig. 10 : Experiment of the control of the luminance intensity

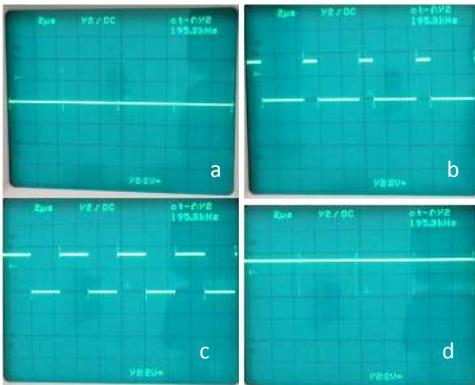


Fig. 11 : oscilloscope visualization (a) for $sw_xx=1$, (b) for $sw_xx=75$, (c) for $sw_xx=128$, (d) for $sw_xx=255$,

Finally, to evaluate improving efficiency of the color appearance, the pixel's color intensity of an RGB LED T-1 3/4 (5mm) FULL COLOR LED LAMP of Kingbright Company [15] was used. The luminance intensity values of each RGB component of this LED were measured by using Lux-meter according to the experimental conditions presented in Table 2. This test was repeated for both method PWM and improved PWM currents.

Table 2 : Experimental conditions.

Parameter	Value
PWM Current reference (I_{Ref})	20 mA
Improved PWM Current reference (I_{NRef})	20 mA
Improved PWM intermediary Current (I_{NRef})	10mA
Temperature	30° C

The result is depicted on Table 3 as a relationship between the number of the pixel under test, between 0 and 255, for the red, green and blue components and the relative intensity of the luminous.

This experimentally measures were illustrated in Fig. 12 as graphs. Based on the interpolation of the practical results presented in Table 3, these graphs show clearly that the improved PWM graphs curves are closely resemble the

conceptualized curves represented in Fig. 6. As expected by design part, the maximum luminance gain occurs in the intermediate area for all three colors. In general, the improved PWM process generates a higher luminance in same PWM average current conditions. For the PWM method, the linear relationship between duty cycles values and luminous intensity has a constant slope. For the improved PWM method, it contains different linear sections in pieces with differentiated slopes.

According to the practical results, average of the experimental values were calculated and presented in Fig. 13. The graphic shows that the Improved PWM are improve up by 12% for Red color, 10% for Green color and 12% for Blue color.

Table 3 : Experimental results of PWM and improved PWM luminance variations

Data	RPW M	RIP WM	GPW M	GIP WM	BPW M	BIP WM
0	0	0	0	0	0	0
25	21	16	49	65	4	5
50	27	43	99	125	7	10
75	49	65	145	180	10	14
100	62	82	175	240	14	19
125	76	98	224	289	18	23
150	81	102	252	311	20	25
175	90	110	300	350	23	28
200	101	118	325	365	25	29
225	110	122	355	380	28	31
250	124	128	370	381	31	32
255	130	131	389	391	32	33

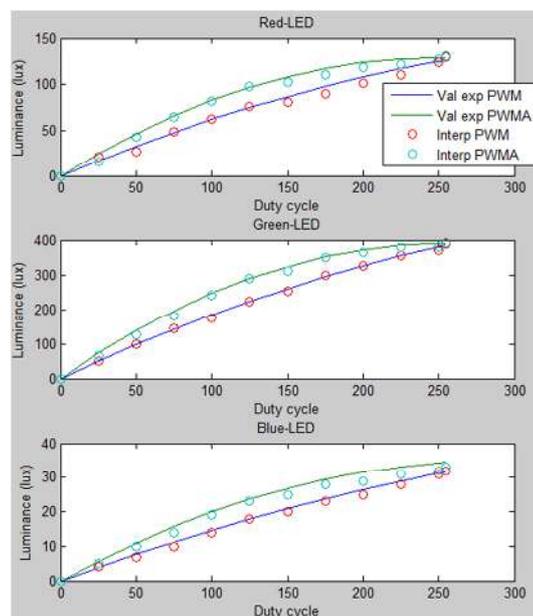


Fig. 12 : Variation of the luminance PWM and MPWM of the R G B LEDs.

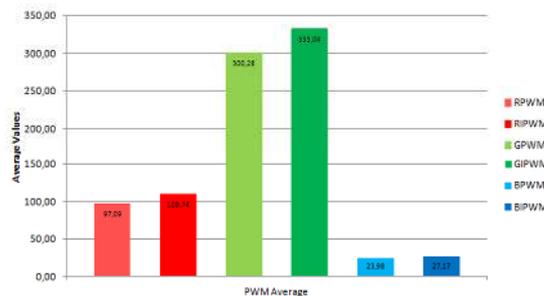


Fig. 13 : Experimental average values of each PWM

5. CONCLUSION

In this paper, the principle of driving LED module display was introduced.

The luminous efficiency of this kind of display deserves further exploration in order to improve the overall efficiency of the system.

Conventional luminance variation methods in these displays (PWM methods) were analyzed and discussed in detail.

The proposed solution to improve PWM method was demonstrated mathematically. It was based on approaching PWM curve to DC curve and reducing the error gap between them. The algorithm for this solution method was proposed using less complex hardware with less changing in module driver structure. This makes it suitable for hardware implementation on an FPGA platform. So, an efficient method for converting pixel's intensity directly to an intensity level of an LED is proposed.

Serial of experiments was realized to check the feasibility of this improved method. Experimental results show that improved PWM can improve the average value by up to 12%, with small addition to the hardware which does not destabilize the operation and structure of the display system.

The luminous efficiency according to the DC variation method is always higher than the improved PWM methods. The Improved PWM curve can reach the DC method curve, using more intermediate points but the hardware becomes more complex and must use more driver modules, which influence in energy consumption and cost of system.

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